

Ch. 13 Transform Coding

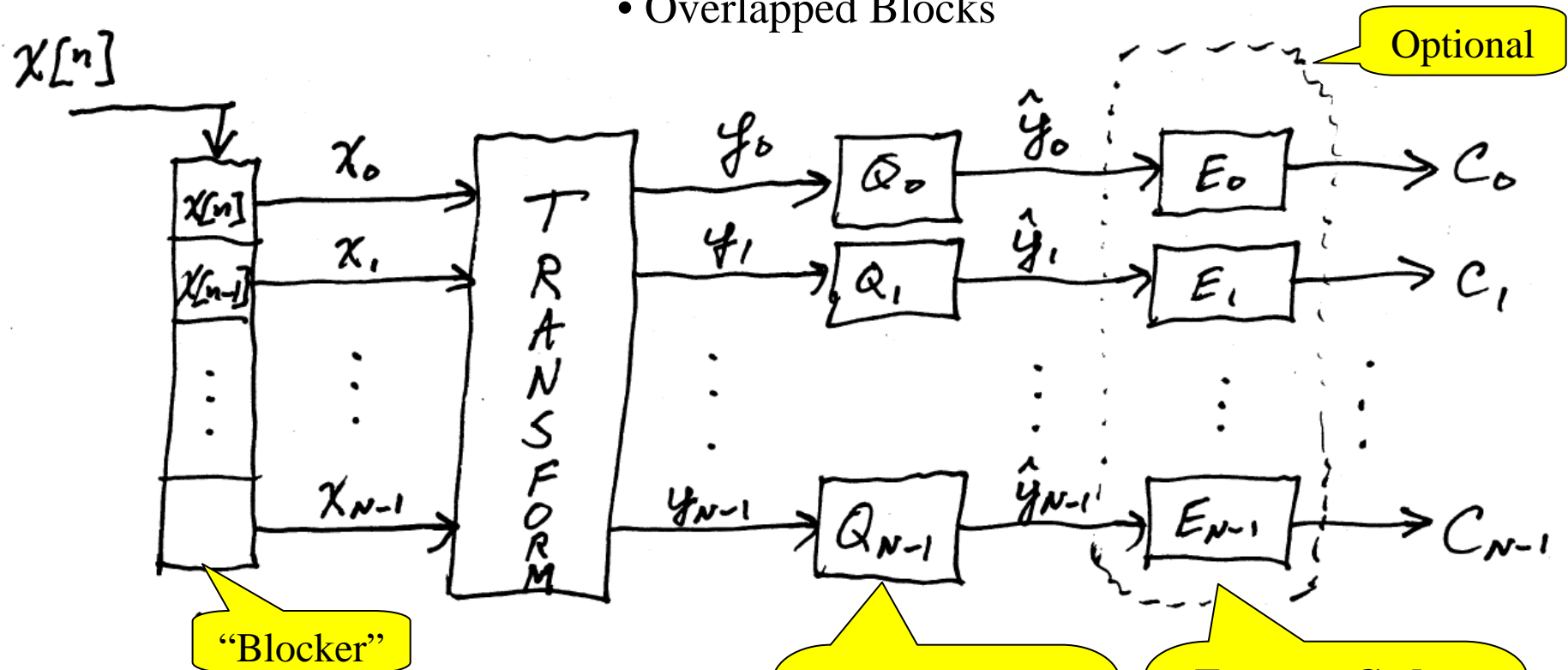
My Coverage is Different from the Book

Overview

Transform the signal (e.g., via the DFT, etc.) into a new domain where compression can be done: (i) better and/or (ii) easier

Often (but not always!) done on a block-by-block basis:

- Non-Overlapped Blocks (most common)
- Overlapped Blocks



**Block Diagram of Transform Coding
"Fig. A"**

Quantizers

Scalar
Vector
Differential

Entropy Coders

Huffmann
Arithmetic
Etc.

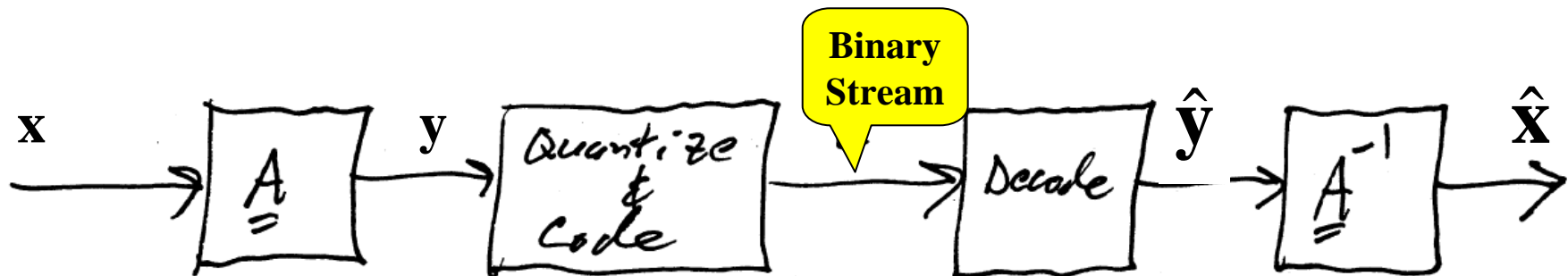
Transform as Linear Operator

We'll view transforms as linear operators on a vector space (finite dimensional):

$$\mathbf{x} = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{N-1} \end{bmatrix} \quad \mathbf{y} = \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_{N-1} \end{bmatrix} \quad \mathbf{x} \xrightarrow{\mathbf{A}} \mathbf{y} \quad \Rightarrow \quad \mathbf{y} = \mathbf{A}\mathbf{x}$$

**A = Operator
... an $N \times N$ Matrix**

Because at the decoder we need to undo the effect of this operator... we need matrix \mathbf{A} to be invertible:



Usefulness of Transform Coding

1. Information Theory Advantages

- Try to make \mathbf{y} have uncorrelated elements
- Try to concentrate energy into just a few elements of \mathbf{y}

2. Perceptual Distortion Advantages

- Transform domain is often better-suited for exploiting aspects of human perception: psychology of hearing and vision

3. Efficient Implementation

- Transform Coding framework provides simple way to achieve #1 & #2
- “Extra” cost of transform is usually not prohibitively large

Need for ON Transforms

Using theory of quantization it is easy to assess transform-domain distortion:

$$d(\mathbf{y}, \hat{\mathbf{y}}) = \frac{1}{N} \sum_{n=0}^{N-1} (y_n - \hat{y}_n)^2$$

But what is the resulting signal-domain distortion?? $d(\mathbf{x}, \hat{\mathbf{x}}) = ???$

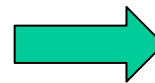
Fact: If transform \mathbf{A} is ON then $d(\mathbf{x}, \hat{\mathbf{x}}) = d(\mathbf{y}, \hat{\mathbf{y}})$

→ Simplifies understanding of impact of quantization choices in the transform domain

Recall: The matrix \mathbf{A} for an ON transform has:

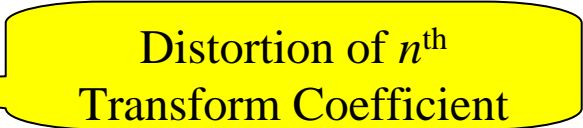
• Columns that are ON vectors: $\mathbf{a}_i^T \mathbf{a}_j = \begin{cases} 1, & i = j \\ 0, & i \neq j \end{cases}$

• Inverse is the transpose: $\mathbf{A}^{-1} = \mathbf{A}^T$



$$\begin{aligned} \mathbf{x} &= \mathbf{A}^T \mathbf{y} \\ \hat{\mathbf{x}} &= \mathbf{A}^T \hat{\mathbf{y}} \end{aligned}$$

So... if the transform is ON then the signal distortion is:

$$\begin{aligned} D &= E \left\{ (\mathbf{x} - \hat{\mathbf{x}})^T (\mathbf{x} - \hat{\mathbf{x}}) \right\} \\ &= E \left\{ (\mathbf{y} - \hat{\mathbf{y}})^T \underbrace{\mathbf{A}\mathbf{A}^T}_{=\mathbf{I}} (\mathbf{y} - \hat{\mathbf{y}}) \right\} \\ &= E \left\{ (\mathbf{y} - \hat{\mathbf{y}})^T (\mathbf{y} - \hat{\mathbf{y}}) \right\} \\ &= E \left\{ \sum_{n=0}^{N-1} (y_n - \hat{y}_n)^2 \right\} \\ &= \sum_{n=0}^{N-1} D_n \end{aligned}$$


Big Picture Result: If ON transform, then Transform-Domain distortions add to give total distortion in signal domain

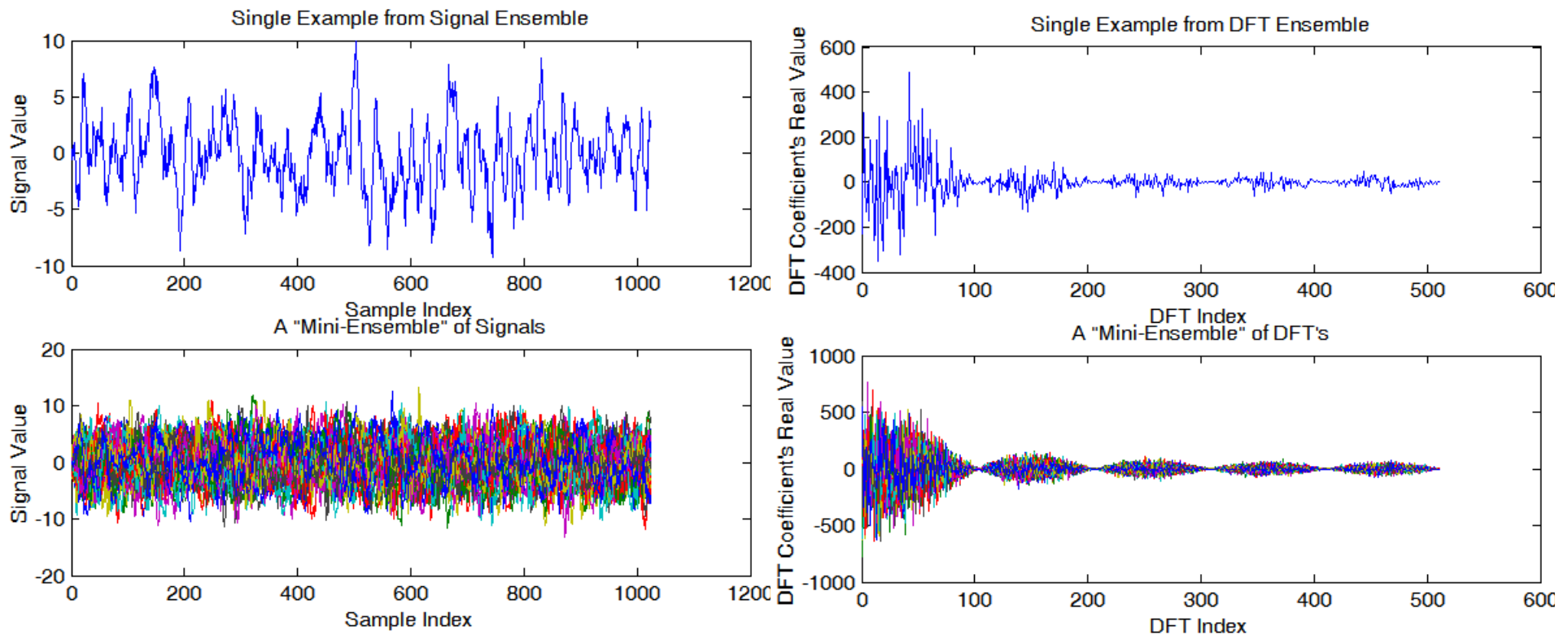
Bit Allocation to TC Quantizers

In “Fig. A” we have N quantizers operating on the transform coefficients...

Q: How do we decide how many bits each of these should use?

This is the so-called “Bit Allocation Problem”... we have a constrained total # of bits... how do we allocate them across the N quantizers?

Q: Why not just allocate them evenly???



Bit Allocation Problem

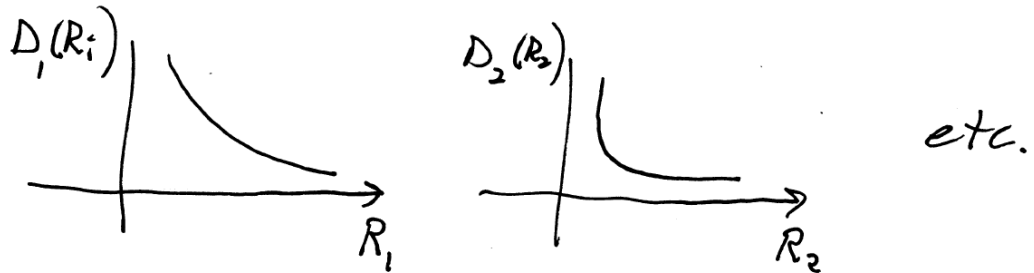
R_B = Total Rate Budget (“Bit Budget”)

R_i = # of bits allocated to the i^{th} quantizer \rightarrow Total Bits Used: $R = \sum_{i=0}^{N-1} R_i$

$D_i(R_i)$ = Distortion of i^{th} quantizer when using R_i bits

Assume distortions are additive (true for ON transform): $D = \sum_{i=0}^{N-1} D_i(R_i)$

Each quantizer has its own R-D curve... depends on quantizer type and char. of the i^{th} transform coefficient



Goal: Allocate bits $\{R_i\}_{i=0}^{N-1}$

to minimize $D = \sum_{i=0}^{N-1} D_i(R_i)$ constrained by $R = \sum_{i=0}^{N-1} R_i \leq R_B$

(Alternate Goal: Minimize R subject to $D \leq D_B$)

Aspects of Bit Allocation

1. Theoretical View
2. Algorithms
 - Average R-D Approach
 - “Operational” R-D Approach

Theory Drives Algorithms

Bit Allocation Theory

Given known functions $D_i(R_i)$

(Based on some appropriate signal & quantizer models)

Models used here determine if we strive for

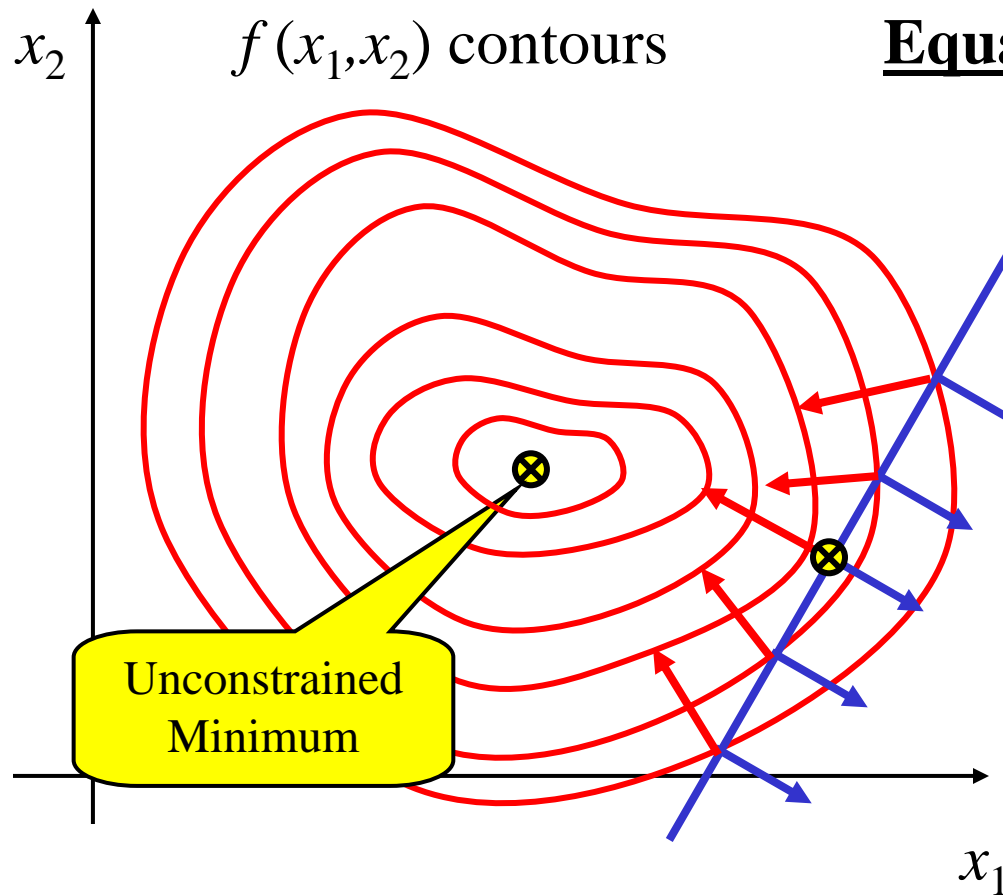
- Average R-D Solution
- Operational R-D Solution

Constrained Opt → Lagrange Mult.

Solve the constrained optimization problem for the optimal allocation vector $\mathbf{r} = [R_0 \ R_1 \ \dots \ R_{N-1}]$

Interpret the result to understand general characteristics

Constrained Optimization: Lagrange Multiplier



Equality Constraint: $g(x_1, x_2) = C$
 $g(x_1, x_2) - C = h(x_1, x_2) = 0$

Ex. $ax_1 + bx_2 - c = 0$
 $\Rightarrow x_2 = (-a/b)x_1 + c/b$
A Linear Constraint

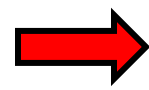
$$\nabla h(x_1, x_2) = \begin{bmatrix} \frac{\partial h(x_1, x_2)}{\partial x_1} \\ \frac{\partial h(x_1, x_2)}{\partial x_2} \end{bmatrix} = \begin{bmatrix} a \\ b \end{bmatrix}$$

Ex. The grad vector has
 “slope” of $b/a \Rightarrow$
 orthogonal to constraint line

Constrained Max occurs when:

$$\nabla f(x_1, x_2) = -\lambda \nabla h(x_1, x_2)$$

$$\Rightarrow \nabla f(x_1, x_2) + \lambda \nabla h(x_1, x_2) = 0$$



$$\nabla_{x_1, x_2, \lambda} \left[f(x_1, x_2) + \lambda (g(x_1, x_2) - C) \right] = 0$$

Lagrange Multiplier Approach to Bit Allocation

See paper: Shoham & Gersho, “Efficient Bit Allocation for an Arbitrary Set of Quantizers,” *IEEE Transactions on Acoustics, Speech and Signal Processing*, Sept. 1988, pp. 1445 – 1453. (*See especially Sect. III*)

Constrained Minimization: $\min_{B \in S} H(B)$ subject to $R(B) \leq R_c$

Theorem: For any $\lambda \geq 0$, the solution $B^*(\lambda)$ to the unconstrained problem

$$\min_{B \in S} \{H(B) + \lambda R(B)\}$$

is also the solution to the constrained problem with constraint

$$\min_{B \in S} H(B) \quad \text{subj. to} \quad R(B) \leq \underbrace{R(B^*(\lambda))}_{\triangleq R^*(\lambda)}$$

So... for each $\lambda \geq 0$ we find the λ -dependent solution to a λ -dependent unconstrained problem... this solution solves a particular version of the constrained problem, where the constraint is λ -dependent

Proof: Since $B^*(\lambda)$ is a solution to the unconstrained problem

$$H(B^*(\lambda)) + \lambda R(B^*(\lambda)) \leq \underbrace{H(B) + \lambda R(B)}_{\forall B \in S}$$

Re-arranging this gives: $H(B^*(\lambda)) - H(B) \leq \lambda [R(B) - R(B^*(\lambda))]$

Since this is true for $B \in S$ it is true for $B \in S^* \subseteq S$ such that $R(B) \leq R(B^*(\lambda))$

$$S^* = \underbrace{\{B \mid R(B) \leq R(B^*)\}}$$

Set of all allocations B that satisfy
 λ -dependent constraint $R(B^*(\lambda))$

Note: $R(B) - R(B^*(\lambda))$ is negative for all $B \in S^*$.

Since λ is positive we have that $H(B^*(\lambda)) - H(B) \leq 0 \Rightarrow H(B^*(\lambda)) \leq H(B)$
 $\forall B \in S^*$

$\Rightarrow H(B^*(\lambda))$ is minimum over all B s.t. $R(B) \leq R(B^*(\lambda))$

$\Rightarrow B^*(\lambda)$ solves the constrained problem with constraint $R(B^*(\lambda))$

<End of Proof>

What does this theorem say?

To each $\lambda \geq 0$...

- there is a constrained problem with: constraint $R^*(\lambda)$
& solution $B^*(\lambda)$ } Both depend on λ
- the unconstrained problem $\min\{H(B)+\lambda R(B)\}$ also has the same solution

So...if we can find closed-forms for $B^*(\lambda)$ & $R^*(\lambda)$ as functions of λ

then we can “adjust” $\lambda = \lambda_c$ so that $R^*(\lambda_c) = R_c$

So we get that... $B^*(\lambda_c)$ solves the constrained problem
w/ our desired constraint

Our Actual
Constraint

Applying the Theorem to TC Bit Allocation

We want to minimize $D(R) = \sum_{i=0}^{N-1} D_i(R_i)$ constrained by $R = \sum_{i=0}^{N-1} R_i \leq R_B$

The theorem says minimize: $J_\lambda = D(R) + \lambda R$ for arbitrary fixed λ
(get results in terms of λ)

$$\Rightarrow J_\lambda = \sum_{i=0}^{N-1} D_i(R_i) + \lambda \sum_{i=0}^{N-1} R_i$$

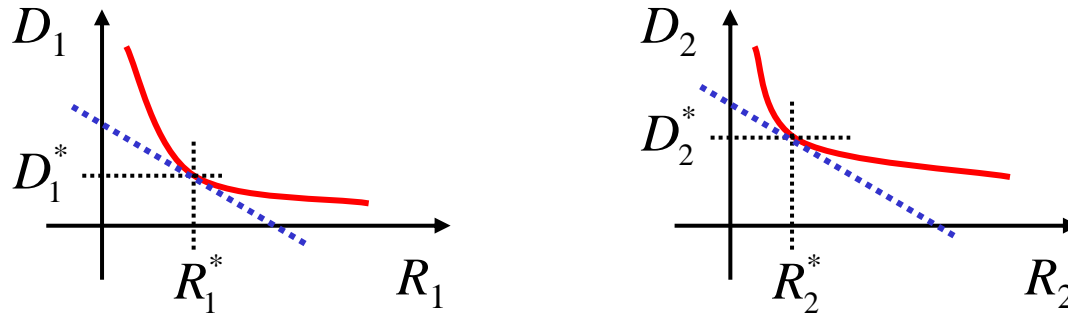
To minimize... we need to set: $\frac{\partial J_\lambda}{\partial R_j} = 0 \quad \forall j \Rightarrow \frac{\partial J_\lambda}{\partial R_j} = \underbrace{\frac{\partial D_j(R_j)}{\partial R_j}}_{\text{set} = 0} + \lambda$

$$\Rightarrow \frac{\partial D_j(R_j)}{\partial R_j} = -\lambda$$

Aha!! Insight!!
→ All the quantizers must operate at an R-D point that has the same slope
“Equal Slopes Requirement”

Intuitive View of “Equal Slopes”

Consider $N = 2$ case



Intuitive “Proof” by Contradiction

1. Assume an optimal operating pt. (R_1^*, D_1^*) & (R_2^*, D_2^*) w/ non-equal slopes

$$S_i^* \triangleq \left. \frac{\partial D_i}{\partial R_i} \right|_{R_i=R_i^*} \quad \text{with } S_1^* \neq S_2^* \quad \text{WLOG: } S_1^* = S_2^* - \Delta \quad \text{w/ } \Delta > 0$$

2. Because assumed optimal: $R_1^* + R_2^* = R_B$ i.e., meets budget
3. Now... imagine small increase in R_1 for quantizer #1: $R_1^* \rightarrow R_1^* + \varepsilon$, $\varepsilon > 0$
4. To keep the bit budget, must decrease rate R_2 by same small amount:

$$R_2^* \rightarrow R_2^* - \varepsilon, \quad \varepsilon > 0 \text{ (same } \varepsilon)$$

5. Find new distortions due to these rate changes... (Use Taylor series approximations... valid because rate changes were small)

$$D_1^* \text{ decreases to } \approx D_1^* + S_1^* \varepsilon = D_1^* + (S_2^* - \Delta) \varepsilon$$

$$D_2^* \text{ increases to } \approx D_2^* - S_2^* \varepsilon$$

6. Find new total distortion:

$$\begin{aligned} D_{New} &= [D_1^* + (S_2^* - \Delta) \varepsilon] + [D_2^* - S_2^* \varepsilon] \\ &= \underbrace{D_1^* + D_2^*}_{D_{Old}} - \underbrace{\Delta \varepsilon}_{> 0} \quad \Rightarrow \quad D_{New} < D_{Old} \end{aligned}$$

Contradiction!!! We assumed we were optimal (but with non-equal slopes)

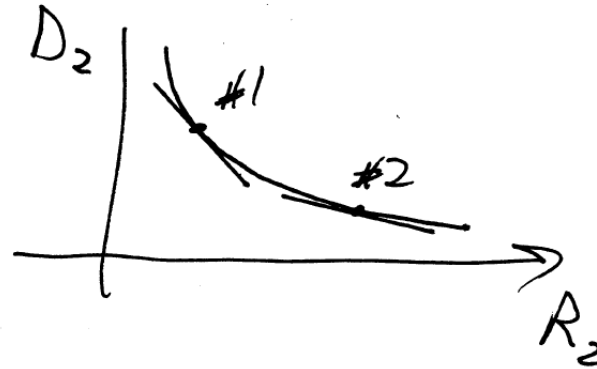
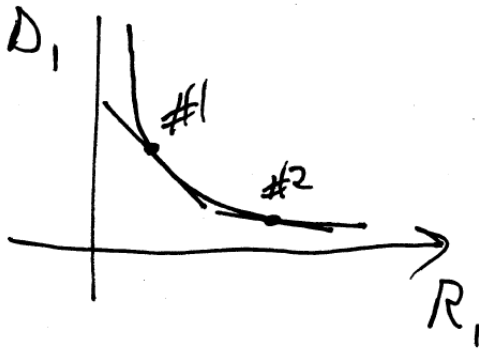
... Yet, we were able to reduce the distortion while meeting bit budget

... So... that non-equal slope operating pt. wasn't optimal after all!!!

... So, equal slopes must occur at the optimal operating point!!!!

So We Need Equal Slopes... But Which Slope?

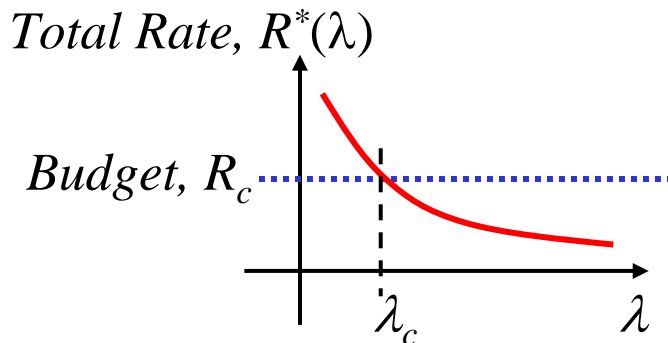
Here are two cases, each with equal slopes... Which should we use?



Note: Slope #1 gives a lower total rate than does Slope #2

→ Choose Slope that causes Total Rate = Budget Rate

Recall: All slopes = $-\lambda$ → **Choose λ to make Total Rate = Budget Rate**



Recall Theorem

Find $\lambda = \lambda_c$ that gives $R^*(\lambda_c) = R_c$
→ Set λ so that the solution to the unconstrained solution also solves our constrained problem with our constraint